EXIT CHART ANALYSIS OF PARALLEL DATA CONVOLUTIONAL CODES

Wei Xiang†
† Faculty of Engineering and Surveying
University of Southern Queensland
Toowoomba, QLD 4350, AUSTRALIA
E-mail: {xiangwei, john}@usq.edu.au

Steven S. Pietrobon†, and John Leis†
† Small World Communications
6 First Avenue
Payneham South SA 5070, AUSTRALIA
E-mail: steven@sworld.com.au

ABSTRACT
We recently proposed a new class of turbo-like codes called parallel data convolutional codes (PDCCs). The distinct characteristics of PDCCs include parallel data input bits and a self-iterative soft-in/soft-out a posteriori probability (APP) decoder. In this paper, we analyse this turbo-like code by means of the extrinsic information transfer chart (EXIT chart). Our results show that the threshold $E_b/N_0$ point for a rate $1/2$ 8-state PDCC is 0.6 dB, which is the same as the threshold point for a punctured rate $1/2$ 16-state parallel concatenated convolutional code (turbo code).

1. INTRODUCTION
The original turbo codes proposed by Berrou et al. [1] are binary turbo codes in that those codes accept only single binary inputs. The so-called non-binary turbo codes are based on a parallel concatenation of RSC component codes with $m$ inputs ($m \geq 2$) [2]. The advantages of non-binary turbo codes include better convergence in iterative decoding, large minimum distances, less sensitivity to puncturing patterns, suboptimum decoding algorithms and reduced latency [2]. Double-binary turbo codes [3] ($m = 2$) usually possess better error-correcting capabilities than binary turbo codes for equivalent implementation complexity and coding rate. This observation led to the use of circular recursive systematic convolutional (CRSC) codes by Berrou et al. [4]. CRSC codes have the advantage of a graceful degradation with increasing coding rate, and are less susceptible to puncturing and suboptimal decoding algorithms. As a consequence, a CRSC code was chosen for the DVB-RCS standard for return channel via satellite [5] as an alternative to concatenated Reed-Solomon (RS) and non-systematic convolutional codes due to their outstanding performance.

Parallel data convolutional codes (PDCCs), a new class of turbo-like codes, were recently proposed by us in [6]. The PDCC encoder inputs are composed of an original block of data and its interleaved version. A novel single self-iterative soft-in/soft-out a posteriori probability (APP) decoder structure is used for the decoding of the PDCCs.

The remainder of this paper is organised as follows. Section 2 briefly reviews the encoding and decoding aspects of PDCCs. Section 3 is dedicated to the EXIT chart analysis of PDCCs, and simulation results are presented in Section 4. Section 5 presents some concluding remarks.

2. PARALLEL DATA CONVOLUTIONAL CODES
In this section, we briefly describe the parallel data convolutional code that we proposed in [6]. The most distinct characteristics of PDCCs include parallel data input bits and a self-iterative soft-in/soft-out a posteriori probability (APP) decoder. Sections 2.1 and 2.2 are dedicated to the encoding and decoding aspects of PDCCs, respectively.

2.1. PDCC Encoder
A new class of turbo-like codes called parallel data convolutional codes (PDCCs) were recently proposed in [6]. Fig. 1 depicts a PDCC encoder in its canonical form which adopts as the constituent convolution code the circular recursive systematic convolutional (CRSC) code proposed for the DVB-RSC standard [5]. The $\pi$ block is an interleaver. It is assumed that $S'_1$ is the MSB (most significant bit) and $S'= 4S'_1 + 2S'_2 + S'_3$.

As depicted in Fig. 1, the block of data sequence to be encoded $A$ and its interleaved version $A'$ constitute two inputs into the encoder. The fact that a PDCC encoder has
two parallel data inputs is the reason that we name it parallel data convolutional codes. X and X' are two systematic outputs, whereas Y and W are two parity bits.

The data stream A and its interleaved version A' are fed into the decoder at the same time. However, A' is decorrelated relative to A due to the presence of the interleaver. For a reasonably good interleaver, like the S-interleaver used in our simulations, this should not adversely affect the performance of the code. The systematic bit X' is not transmitted as X' is the interleaved version of X. Thus, the PDCC encoder shown in Fig. 1 can typically provide a code rate of 1/2 by transmitting the systematic bit X and the parity bit Y, and a code rate of 1/3 by transmitting the systematic bit X and the parity bits Y and W. It can also provide other coding rates through puncturing the parity bits Y and W if needed.

2.2. Self-iterative PDCC Decoder

The key difference between the MAP algorithm for PDCCs and the MAP algorithm presented in [7] is that the PDCC encoder has two input bits and four output bits, including two systematic bits A, A' and two parity bits Y, W. The MAP algorithm described in [7], however, is applicable to the soft decoding of rate 1/2 systematic convolution codes which have one input bit and two output bits, including one systematic bit and one parity bit.

Assume the outputs of the PDCC encoder depicted in Fig. 1 at time index k are the systematic bit A_k, and the parity bits Y_k and W_k. These outputs are BPSK modulated and transmitted through an AWGN channel. At the receiver end, the received symbols at time index k are denoted as R_k, R_Y_k, and R_W_k, respectively. A'_k, the interleaved version of the received symbol A_k, is obtained by interleaving A_k at the receiver, and thus will not be transmitted. It can be shown that the branch metric γ_k^{i,m} which denotes the branch exiting from S_k = m with A_k = i, can be expressed as

\[ \gamma_k^{i,m} = \chi_k \xi_k^i \xi_k^j \exp \left( -L_c (R_A_k + R_A'_k A_k' + R_Y_k Y_k + R_W_k W_k) \right) \]

where \( \chi_k \) is a constant, \( \xi_k^i = \Pr(A_k = i) \), \( \xi_k^j = \Pr(A_k' = j) \), and \( L_c = 2/\sigma^2 \). The likelihood ratio \( \lambda_k \) associated with each decoded bit \( A_k \) is compared to a threshold equal to one in order to determine the decoded bit \( A_k \).

The novelty of decoding the PDCCs lies in self-iterative decoding. The self-iterative PDCC decoder operates like a normal MAP decoder except it feeds the extrinsic outputs after interleaving or deinterleaving back as a priori inputs. Fig. 2 shows a schematic of a self-iterative PDCC decoder.

The inputs to the decoder are the soft outputs of a noisy channel \( L_c R_A \), \( L_c R_Y \) and \( L_c R_W \), respectively. The decoder reconstructs \( L_c R'_A \) by interleaving \( L_c R_A \). The idea of self-iterative decoding comes from the fact that \( R'_A \) is the interleaved version of \( R_A \), so that the extrinsic information of \( R_A \) can be fed back as the a priori information for \( R'_A \) after interleaving and the extrinsic information of \( R'_A \) can be fed back as the a priori information for \( R_A \) after deinterleaving.

We denote the a priori information of \( R_A \) and \( R'_A \) by \( Z_A^r \) and \( Z'_A \), while the extrinsic information of \( R_A \) and \( R'_A \) are denoted by \( Z_A^e \) and \( Z'_A \), respectively. The self-iterative MAP decoder computes the APP of the information bit \( A \). The LLR output of the decoder can be expressed as

\[ L_{out} = L_c R_A + Z_A^r + Z_A^e + Z'_A + Z'_A. \]  

The self-iterative PDCC decoder proceeds as follows. At the first decoding iteration, \( Z_A^r = Z'_A \) are initialised to zero. For the subsequent iterations, \( Z_A^r \) is interleaved and fed back as the a priori information for \( A' \), i.e., \( Z_A^r = \pi(Z_A^r) \) where \( \pi(\cdot) \) denotes an interleaving mapping. Likewise, \( Z'_A \) is deinterleaved and fed back as the a priori information for \( A \), i.e., \( Z_A^e = \pi^{-1}(Z'_A) \) where \( \pi^{-1}(\cdot) \) denotes a deinterleaving mapping. At the final iteration, the decoder delivers the log-likelihood output \( L_{out} \). The self-iterative decoding process can be clearly seen from the two feedback connections between \( Z_A^r \) and \( Z'_A \), \( Z'_A \) and \( Z_A^e \) in Fig. 2.

3. EXIT CHART ANALYSIS OF PDCCS

The extrinsic information transfer chart (EXIT chart) [8] is a powerful tool for analysing the convergence behavior of iterative decoding of turbo-like codes. The essential idea of the EXIT chart lies in the fact that it can predict the behavior of an iterative decoder by looking solely at the input/output relations of individual constituent decoders. The EXIT chart analyses the input/output characteristics of a single soft-input/soft-output (SISO) decoder by observing the extrinsic information at the output of the decoder for a
range of a priori input. It then uses mutual information to describe the extrinsic information transfer characteristics of an iterative SISO decoder.

The EXIT chart analysis is based on two empirical observations obtained by simulation. First, the a priori information $A$ remains uncorrelated from the channel observations $Z$ for large interleavers. Second, the extrinsic output $E$ yielded by one constituent decoder approaches a Gaussian-like distribution with increasing number of iterations.

As discussed in [8], the a priori information $A$ is measured in terms of mutual information $I_A = I(X; A)$ between the transmitted systematic information bits $X$ and $A$ in $L$-values [9] as

$$I_A = \frac{1}{2} \sum_{x=-1,1} \int_{-\infty}^{\infty} p_A(\xi \mid X = x) \cdot \log_2 \left( \frac{2p_A(\xi \mid X = x)}{p_A(\xi \mid X = -1) + p_A(\xi \mid X = 1)} \right) d\xi. \quad (3)$$

Similarly, the extrinsic output $E$ of the SISO decoder can also be measured in terms of mutual information $I_E = I(X; E)$ between the transmitted systematic information bits $X$ and the extrinsic information $E$ in $L$-values as

$$I_E = \frac{1}{2} \sum_{x=-1,1} \int_{-\infty}^{\infty} p_E(\xi \mid X = x) \cdot \log_2 \left( \frac{2p_E(\xi \mid X = x)}{p_E(\xi \mid X = -1) + p_E(\xi \mid X = 1)} \right) d\xi. \quad (4)$$

The convergence behaviour of the iterative decoder can be described as a mapping between mutual information $I_A$ and $I_E$.

In order to investigate the convergence behaviour of the self-iterative PDCC decoder depicted in Fig. 2, we apply the EXIT chart algorithm to PDCCs in this paper. The fundamental difference between the PDCC EXIT chart analysis and the parallel concatenated convolutional codes (PCCCs or turbo codes) EXIT chart analysis lies in the fact that generating PCCC EXIT charts does not need an interleaver, while generating PDCC EXIT charts does need an interleaver. This is because the self-iterative PDCC decoder has two received systematic channel inputs in parallel, with one systematic channel input $L_C R_A$ being the interleaved version of the other systematic channel input $L_C R_A$ as shown in Fig. 2. As a result, we need to prepare two a priori inputs to the self-iterative PDCC decoder for applying the EXIT chart algorithm to the PDCC.

Assume $x$ and $x'$ are the two systematic information data inputs of the PDCC encoder. In the EXIT chart analysis, the two a priori inputs $A$ and $A'$ to the PDCC decoder corresponding to the two information data inputs $x$ and $x'$ can be modeled as follows

$$A_x = \mu_A \cdot x + n_A \quad (5)$$
$$A_{x'} = \pi(A_x) \quad (6)$$

where $n_A$ is an independent Gaussian random variable with variance $\sigma^2_A$ and zero mean, $\mu_A = \sigma^2_A / 2$, and $\pi(\cdot)$ denotes an interleaving function. Equation (6) implies that we could use an interleaver to interleave the a priori input for $x$ to yield the a priori input for $x'$.

4. SIMULATION RESULTS

In this section, we compare the performance of the PDCCs and PCCCs by means of EXIT chart analysis.

The bit error rate (BER) performance comparison between PDCCs and PCCCs was presented in [6]. The simulation configurations were that both the PDCC and PCCC have coding rate 1/2 and a block size of 8192 random information bits. An $S$-type interleaver [10] with $S$ equal to 47 was used. It was shown that the performance of PDCCs is about 0.2 dB inferior to that of PCCCs at low BERs, although the performance difference of the two codes was negligible for low $E_b/N_0$ up to 0.6 dB.

The relatively inferior performance of PDCCs was diagnosed to be caused by the so-called “self-terminating” phenomena of the PDCC. For the PCCC, an error bit could cause the trellis path to divert from the two all-zero paths. The same bit is interleaved and fed into the second constituent encoder. That bit would not cause the diverted trellis path to re-emerge earlier. On the other hand, for the PDCC, an error causes a diversion from the all-zero trellis path. The same bit is interleaved and then fed into the same PDCC encoder. That bit could cause an earlier trellis remerge and thus self-terminating.

The PDCC performance using the BER measurement is largely dependent on the interleaver structure and size. However, the EXIT chart analysis will tell us the minimum $E_b/N_0$ that can be achieved with an infinite size interleaver and infinite iterations. Fig. 3 graphically shows the EXIT charts for the PCCC at various $E_b/N_0$ values, whereas Fig. 4 presents the EXIT charts for the PDCC. The PCCC used in our simulation is the original punctured rate 1/2 16-state turbo code with forward and backward polynomials (21, 37) in octal [1]. For the PCCC EXIT charts, the block size is 65536 and no interleaver is used. For the PDCC EXIT charts, an $S$-type interleaver with $S$ equal to 192 is used and the block size is also 65536.

As can be seen from Figs. 3 and 4, the $E_b/N_0$ threshold for the PCCC is around 0.6 dB, whereas the $E_b/N_0$ threshold for the PDCC is also around 0.6 dB. Therefore, for an infinite size interleaver and infinite iterations, the EXIT chart analysis indicates that the performance of the PDCC is comparable to that of the PCCC, although the BER performance of the PDCC presented in [6] is inferior to that of the PCCC due to the “self-terminating” property of the PDCC. Future research will examine PCCCs with the constituent code in Fig. 1, to allow a fairer comparison with the code used for
the PDCC.

![PDCC EXIT charts with an block size of 65536.](image1)

![PDCC EXIT charts with an interleaver size of 65536.](image2)

**5. CONCLUSIONS**

In this paper, the results using the EXIT chart analysis applied to PDCCs are presented. The PDCC features a parallel interleaved systematic data input and a self-iterative decoder.

Previous results by means of Monte Carlo BER simulation showed that PDCCs may have inferior performance compared to PCCCs. This is due to the PDCC’s undesirable self-terminating property. Therefore, the interleaver structure and its robustness to the self-terminating phenomena has a strong influence on the performance of PDCCs using Monte Carlo simulation. However, the EXIT chart analysis results presented in this paper reveal that the performance of the PDCC is close to that of the PCCC. Future research in this area includes designing self-terminating resilient interleavers to push the PDCC performance using iterative decoding close to its theoretic limit revealed by the EXIT chart analysis.

**6. REFERENCES**


